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# Beamforming Performance Analysis for OFDM Based IEEE 802.11ad Millimeter-Wave WPANs

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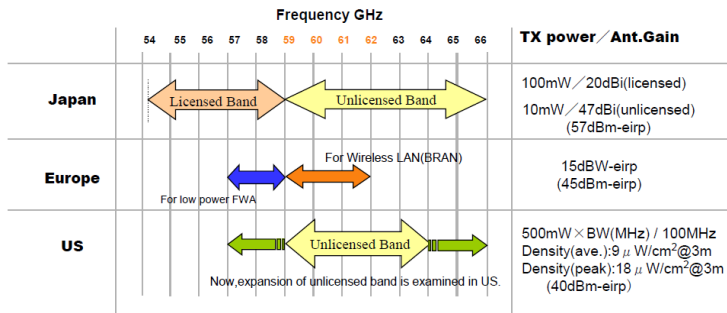
8<sup>th</sup> International Workshop on  
Multi-Carrier Systems and Solutions

# Outline

- 1 Introduction
  - Overview of Wireless Personal Area Network (WPAN)
  - IEEE 802.11ad Standard
- 2 System Model
  - Channel Frequency Response
  - Optimization Criteria
- 3 OFDM Based Beamforming
  - Subcarrier-wise
  - Symbol-wise
  - Hybrid
- 4 Numerical Results
  - Beamforming Gain
  - BER Performance
  - Link Throughput and Ranges

## Overview of Wireless Personal Area Network (WPAN)

## 60 GHz Frequency Band Allocation



- Large availability of 7 GHz unlicensed in worldwide
- Potentially small device components

# Overview of 60 GHz WPAN

## Standards over 60 GHz WPAN

- IEEE 802.15.3c
- WirelessHD
- WiGig
- ECMA-387
- **IEEE 802.11ad**

## Characteristics of 60 GHz millimeter-wave WPANs

- In-door (<10m)
- Uncompressed HDTV and high rate data transfer
- At least 1 Gbps throughput, 3-4 Gbps preferable

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# Operating Modes

- Single Carrier: Low complexity and control information
- OFDM: High performance applications

Table: OFDM Modulation and Coding Schemes

Modulation	Coding Rate	Coded Bits/Symbol	Data Bits/Symbol	Data Rate (Mbps)
QPSK	1/2	672	336	1386.00
QPSK	5/8	672	420	1732.50
QPSK	3/4	672	504	2079.00
16-QAM	1/2	1344	672	2772.00
16-QAM	5/8	1344	840	3465.00
16-QAM	3/4	1344	1008	4158.00
16-QAM	13/16	1344	1092	4504.50
64-QAM	5/8	2016	1260	5197.50
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# MIMO Communication System

Let  $y_m$  be the received decision baseband signal for the  $m$ th subcarrier

$$y_m = \tilde{H}_m x_m + n_m, \quad m = 1, \dots, N$$

where  $\mathbf{x}_m$  is the transmitted data symbol,  $\mathbf{n}_m$  is the Gaussian noise vector with zero mean and variance  $\sigma^2$ ,  $N$  is the number of subcarriers, and  $\tilde{H}_m$  represents the frequency response of the equivalent channel matrix for the  $m$ th subcarrier after beamforming, which is given by:

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# Choose the Optimal Weight Vectors

## Optimization Criteria

- Max-codeword-distance
- Max-BER
- Max-mutual-information
- **Max-effective-SNR**

$$\gamma_{\text{eff}} = -\beta \ln \left[ \frac{1}{N} \sum_{m=1}^N \exp(-\gamma_m/\beta) \right]$$

where  $\gamma_m$  is the symbol SNR experienced on the  $m$ th subcarrier,  $\beta$  is a parameter dependent on MCS.

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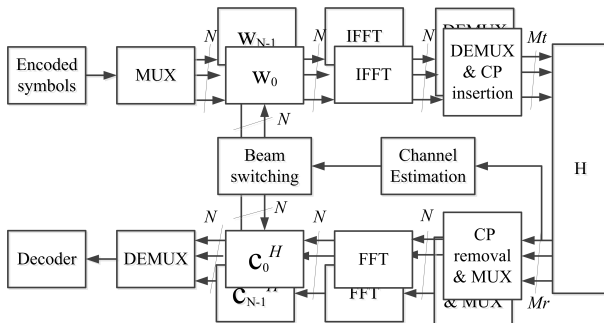
$$\gamma_m = \frac{E \left[ |c^H H_m w x_m|^2 \right]}{E \left[ |n_m|^2 \right]} = \frac{|c^H H_m w x_m|^2}{M_t M_r \sigma^2}$$

where  $M_t$  and  $M_r$  are the number of antenna elements at the transmitter and the receiver respectively. When normalized,  $w^H w = M_t$  and  $c^H c = M_r$ .

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# Maximize the Received SNR on Each Subcarrier



**Figure:** Block diagram of subcarrier-wise beamforming

$$\gamma_{\text{eff}, \text{subcarrier}} = -\beta \ln \left[ \frac{1}{N} \sum_{m=1}^N \exp \left( -\frac{\max_{c,w} |c^H H_m w|^2}{\beta M_t M_r \sigma^2} \right) \right]$$

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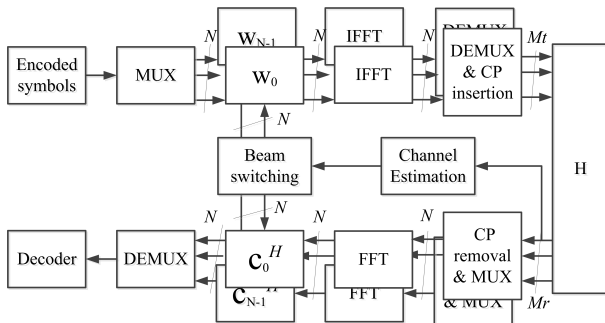


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## Subcarrier-wise

# Maximize the Received SNR on Each Subcarrier

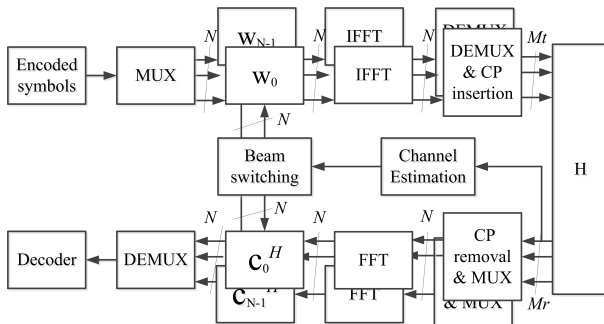


Figure: Block diagram of subcarrier-wise beamforming

Optimal but not practical

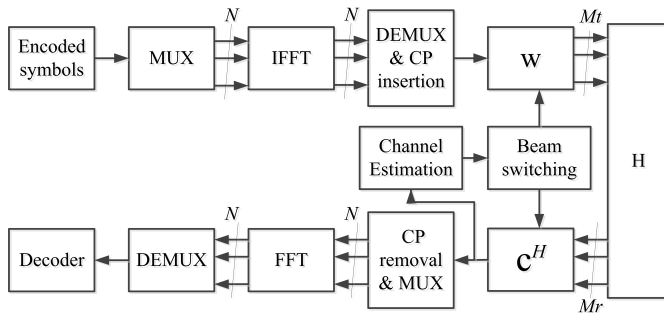
- Need full channel state information
- Requires one FFT/IFFT processor per antenna

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## Symbol-wise

# Each Subcarrier Applies the Same Weight Vector



**Figure:** Block diagram of symbol-wise beamforming

## Pre-defined beam codebook

- Full channel state information is not required
- Depends on the number of antenna elements and beams

## Symbol-wise

# Each Subcarrier Applies the Same Weight Vector

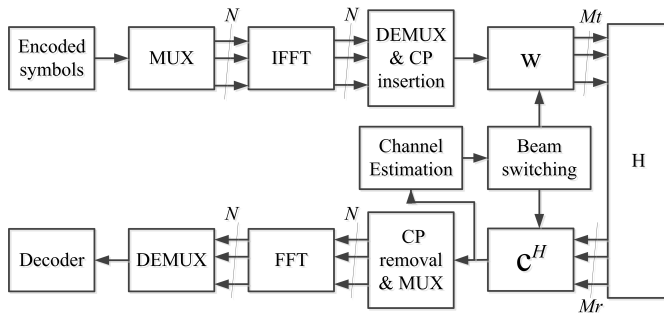


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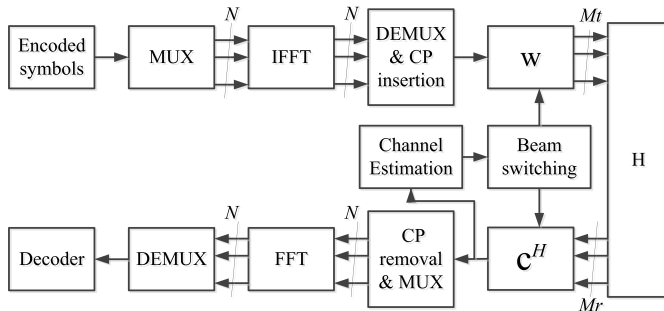


Figure: Block diagram of symbol-wise beamforming

$$\gamma_{eff,symbol} = \max_{c,w \in C} (-\beta) \ln \left[ \frac{1}{N} \sum_{m=1}^N \exp \left( -\frac{|c^H H_m w|^2}{\beta M_t M_r \sigma^2} \right) \right]$$

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# Compromise the complexity and performance

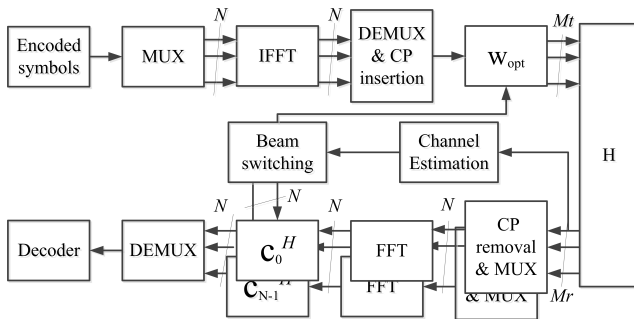


Figure: Block diagram of hybrid beamforming

Symbol-wise at Tx, and subcarrier-wise at Rx

- Optimal each receiver steering vector
- Also use pre-defined codebook

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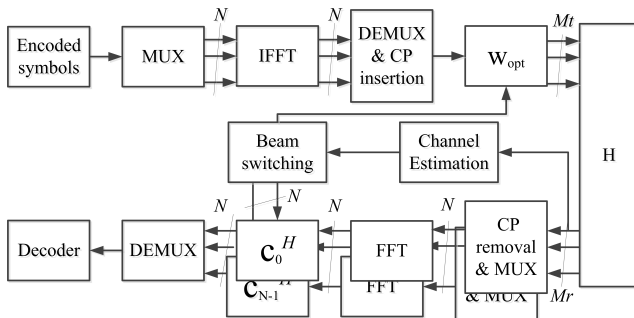


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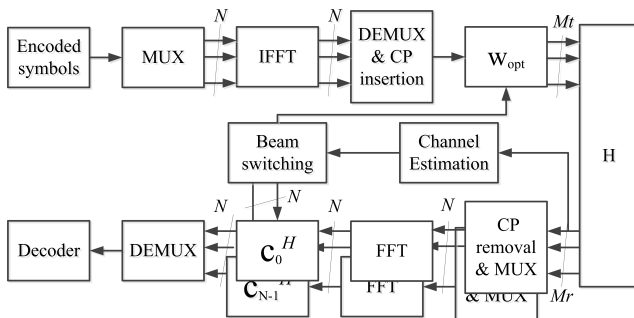


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$$\gamma_{eff, hybrid} = \max_{w \in C} (-\beta) \ln \left[ \frac{1}{N} \sum_{m=1}^N \exp \left( -\frac{|C^H H_m w_{opt}|^2}{\beta M_t M_r \sigma^2} \right) \right]$$

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# Preliminaries

## System assumptions

- $N=512$  OFDM subcarriers
- 1D half wavelength isotropic radiators
- $M= M_t = M_r$  antenna elements

## Channel assumptions

- 60 GHz channel models
- Both LOS and NLOS

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# LOS Scenario

## Evaluate beamforming performance

$$G = \frac{\gamma_{\text{eff}, \text{beamforming}}}{\gamma_{\text{eff}, \text{SISO}}}$$

- Beamforming gain has a bound when single path exists
- The performance difference is not noticeable, because the LOS component exists

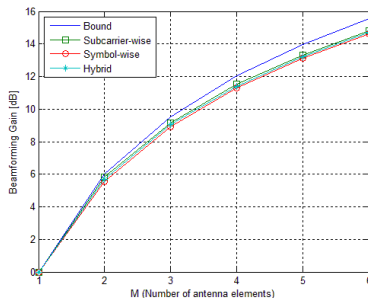


Figure: Beamforming gain with LOS

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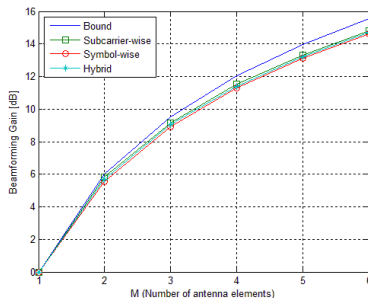


Figure: Beamforming gain with LOS

# NLOS Scenario

## Evaluate beamforming performance

$$G = \frac{\gamma_{\text{eff}, \text{beamforming}}}{\gamma_{\text{eff}, \text{SISO}}}$$

- Subcarrier-wise is the best, hybrid is the next and symbol-wise is the worst
- The more antenna elements, the higher improvement can be achieved by hybrid beamforming

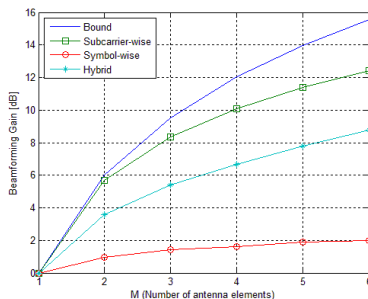


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# Bit Error Rate

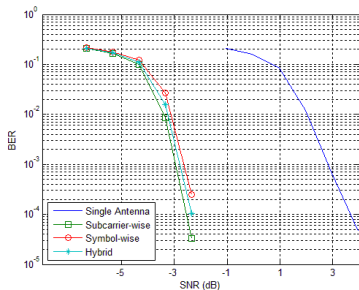


Figure: BER for QPSK 1/2 with LOS

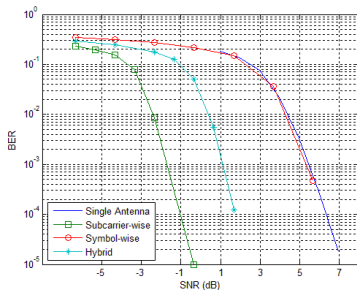


Figure: BER for QPSK 1/2 with NLOS

- A 2-by-2 antenna system is assumed
- The simulated BER performance verified the numerical results

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# Link Throughput in LOS

## Link Adaptation Scheme

- The PHY mode with highest throughput will be selected:

$$\text{Throughput} = R(1 - \text{PER})$$

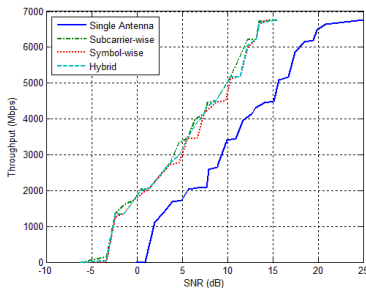


Figure: Link throughput with LOS

- The throughput envelop is the ideal adaptive MCS based on the optimum switching point
- At a certain SNR, beamforming systems offer higher throughput than SISO

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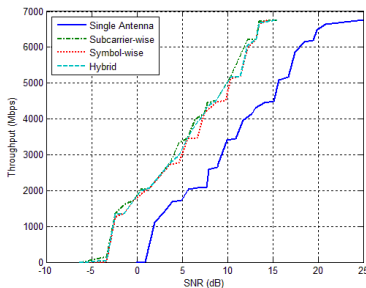


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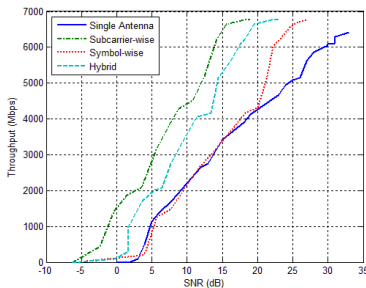


# Link Throughput in NLOS

## Link Adaptation Scheme

- The PHY mode with highest throughput will be selected:

$$\text{Throughput} = R(1 - \text{PER})$$



- Beamforming schemes do not improve the peak error-free throughput
- More gain can be achieved for very high throughput (>4500 Mbps)

Figure: Link throughput with NLOS

# Operation Range in LOS

## Path Loss Model

$$PL(dB) = A + 20 \log_{10}(f) + 10n \log_{10}(D)$$

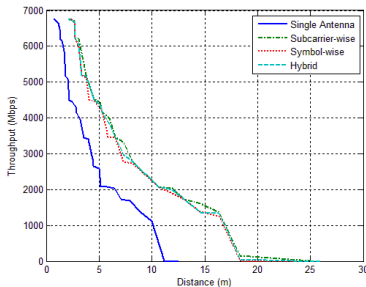


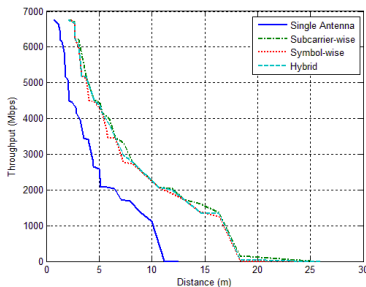
Figure: Operation range in LOS

- The system operates at its maximum throughput when the device are close
- Adaptively switch to the lower speed when a device moves further away

# Operation Range in LOS

## Link Budget Model

$$P_T - PL \geq kTB + NF + \text{ReceiverSNR}$$



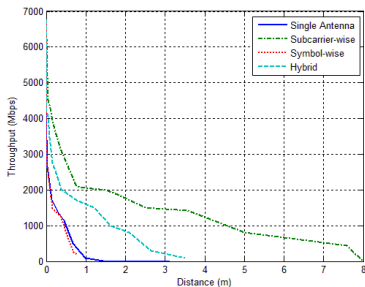
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Figure: Operation range in LOS

# Operation Range in NLOS

## Link Budget Model

$$P_T - PL \geq kTB + NF + \text{ReceiverSNR}$$



- The SISO system could not provide service beyond 1m
- Subcarrier-wise and hybrid beamforming extend the achievable range significantly

Figure: Operation range in NLOS

# Summary

- A performance evaluation of three types of beamforming schemes over the OFDM based 60 GHz WPAN are studied;
- Beamforming schemes increase the system performance significantly;
- In NLOS, hybrid beamforming provide considerable improvements while maintaining reasonable hardware complexity

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# For Further Reading I



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Hybrid beam-forming and beam-switch for OFDM based WPAN

*JSAC*, 27(8):1425-1432, Oct 2009.



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A. Maltsev, *et.al*

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May 2010.



# Thank you! and Questions?

or Email to <x.zhu@Bristol.ac.uk>